Meeting S3, Paris, CNAM



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Tolérance aux fautes par une approche superviseur: gestion du couplage FDI-FTC

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Outline



- Preliminary
- A brief state-of-art
 - Passive approaches
 - Active approaches
 - Closing words -> Limitations

• Supervisory FTC with mutual perf. optim.

- Theory and principles
- Overall stability
- Optimization issue
- Academic illustration
- Closing words

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Preliminary

State-Of-Art

•Example

Supervisory FTC

Closing words





Preliminary 1/2



• Fault Tolerant Control (FTC) strategies are meant to manage faulty situations by maintaining overall system stability and acceptable performances.

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- **Fault tolerant control** (FTC) deals with a concept for handling faulty situations by suitable reconfiguration of the control laws
- It is fundamentally **a control problem** (and nothing else)
 - at least, one fault occurs in the system (of course, these faults have to be diagnosed)
 - performances achieved by the already in place control laws <u>are no more satisfactory</u>

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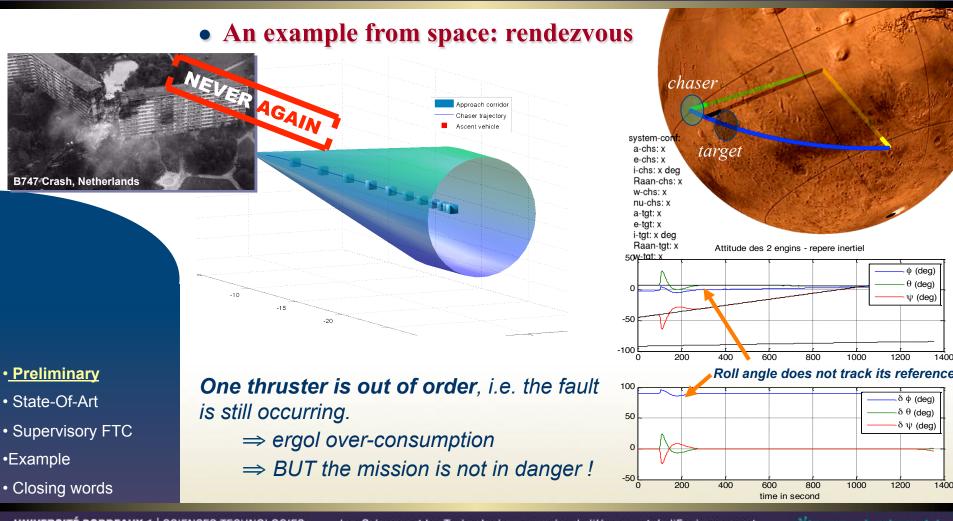




Preliminary 2/2



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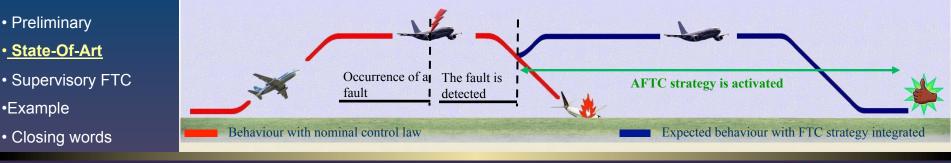
State-of-art 1/4



- **Two main approaches:** Different controller architectures for FTC have been suggested in the literature, see (Aström K et al, 2000; Zhang & Jiang, 2008; Noura et al, 2009 for good surveys)
 - Passive Fault-Tolerant Control systems (PFTC):
 - nothing else than robust control approaches against pre-specified faults
 - has limited fault-tolerant capabilities

• Active Fault-Tolerant Control systems (AFTC):

- no degradation of fault-free operating mode
- principle: 1) Detect the fault event (FDI issue);
 - 2) Activate the fault compensation mechanism (switching logic usually)
 - 3) Reconfigure the control laws (FTC issue) or mission objectives (FTG issue)



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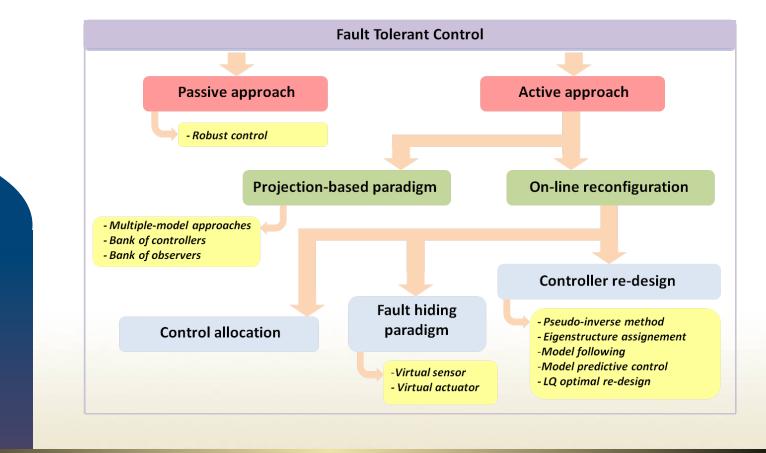
•Example



State-of-art 2/4



• **Classification:** Inspired by Lunze et al 2006 and Zhang & Jiang 2008



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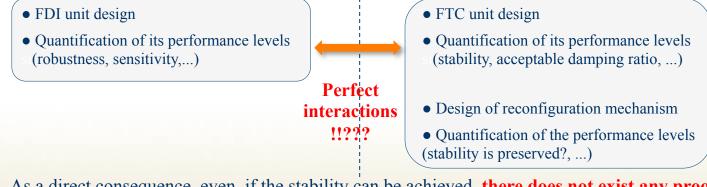
State-of-art 3/4



Closing words of State-of-art: AFTC limitations



• **Open problem:** Guaranteeing stability and performances of the overall fault tolerant scheme taking into account the FDI, the switching and the re-configuration mechanisms, is not considered. In practice, coupling properties are studied by means of a Monte-Carlo campaign.



 Supervisory FTC As a direct consequence, even if the stability can be achieved, there does not exist any proof of global optimality of the FTC scheme since the controllers and the FDI/fault estimation algorithms are designed separately.



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•Example

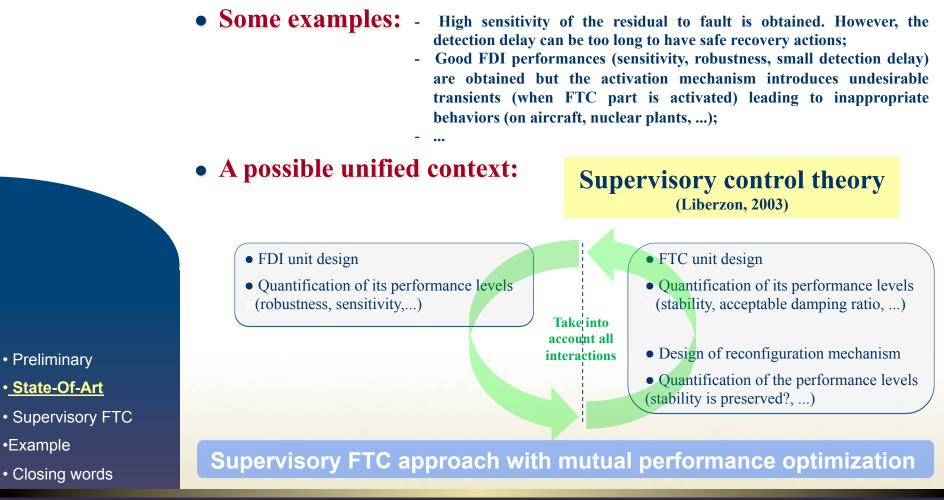
Preliminary

State-Of-Art



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Supervisory FTC

Supervisory FTC with mutual perf. 1/11



• Objectives:

- Formal stability proofs are established for the overall FTC scheme taking into account the plant model switching, the control reconfiguration switching and the influence of uncertainties and unknown inputs.
- The method allows to design both the FDI and FTC units taking into account their coupling. The method allows to derive a FDI and fault tolerant controller scheme with guaranteed stability and well established performance in terms of robustness, fault detection and tolerance.
- It is proved that the global **stability of the control law is preserved even if the FDI scheme fails** to identify the correct fault. In this case, there may exist a system chattering effect that can be reduced by choosing some adequate parameters.

A great advantage for space missions since it is not necessary to switch off the diagnosis and the tolerance algorithms, global stability of the GNC being formally proved in both fault free and faulty situations.

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Meeting S3, Paris, CNAM - January, 18, 2013



Supervisory FTC with mutual perf. 2/11



- **Theory & principles:** D. Efimov, J. Cieslak, D. Henry, Supervisory fault-tolerant control with mutual performance optimization, *Int. J. Adapt. Control & Signal Proc.*, 2012
 - System model: $\mathbf{\hat{x}}_{p} = \mathbf{A}_{p}\mathbf{x}_{p} + \mathbf{B}_{p}\mathbf{u}_{p} + \mathbf{G}_{p}\mathbf{d}$ $\mathbf{y}_{p} = \mathbf{C}_{p}\mathbf{x}_{p}$

- Assume a already in-place nominal control:

 $\mathbf{x}_{c} = \mathbf{A}_{c}\mathbf{x}_{c} + \mathbf{B}_{c}\mathbf{y}_{p}$ $\mathbf{y}_{c} = \mathbf{C}_{c}\mathbf{x}_{c} + \mathbf{D}_{c}\mathbf{y}_{p}$

Then, considering actuator and component faults:

$$\mathbf{\hat{x}}_{p} = (\mathbf{A}_{p} + \Delta \mathbf{A}_{i})\mathbf{x}_{p} + (\mathbf{B}_{p} + \Delta \mathbf{B}_{i})\mathbf{u}_{p} + \Delta_{i} + \mathbf{G}_{p}\mathbf{d} \qquad i = \overline{1, N}$$
$$\mathbf{y}_{p} = \mathbf{C}_{p}\mathbf{x}_{p}$$

- matrices ΔA_i , ΔB_i multiplicative faults
- vectors Δ_i additive faults.

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• <u>Objective:</u>

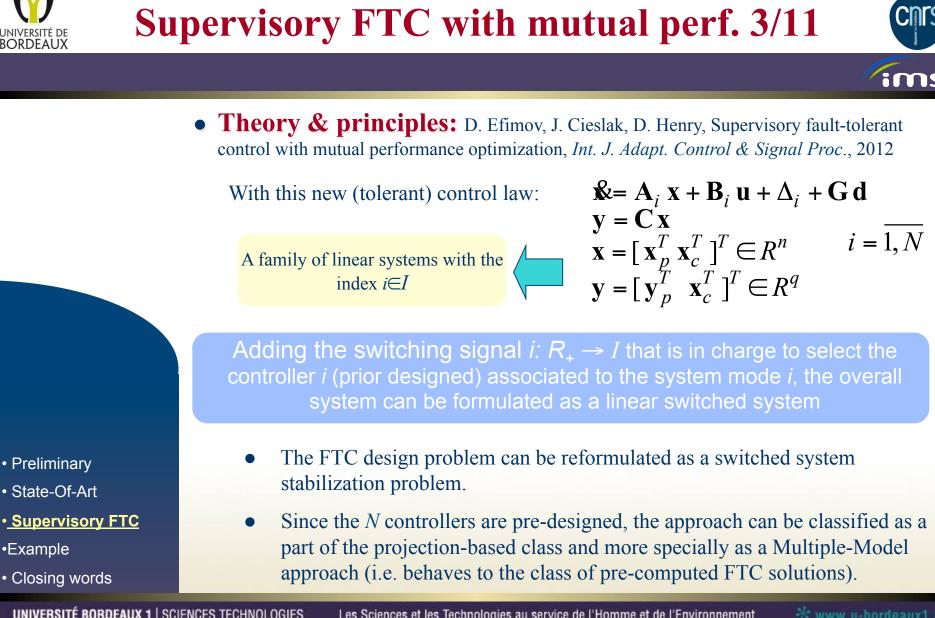
The goal is to design the control signal u in parallel with the nominal controller (A_c, B_c, C_c, D_c) so that $\mathbf{u}_p = \mathbf{y}_c + \mathbf{u}$

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•Example Closing words

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Supervisory FTC with mutual perf. 4/11



• Theory & principles:

- The problem of supervisory FTC design has already been addressed in the literature (Blanke, *et al.*, 1997; 2003; Boskovic and Mehra, 2002)
- Many approaches have been applied for <u>independent</u> optimization of the fault detection, isolation and compensation systems.

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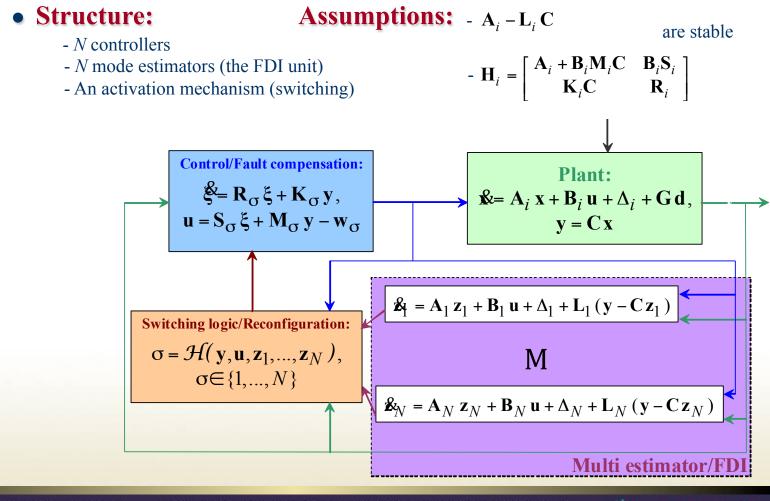






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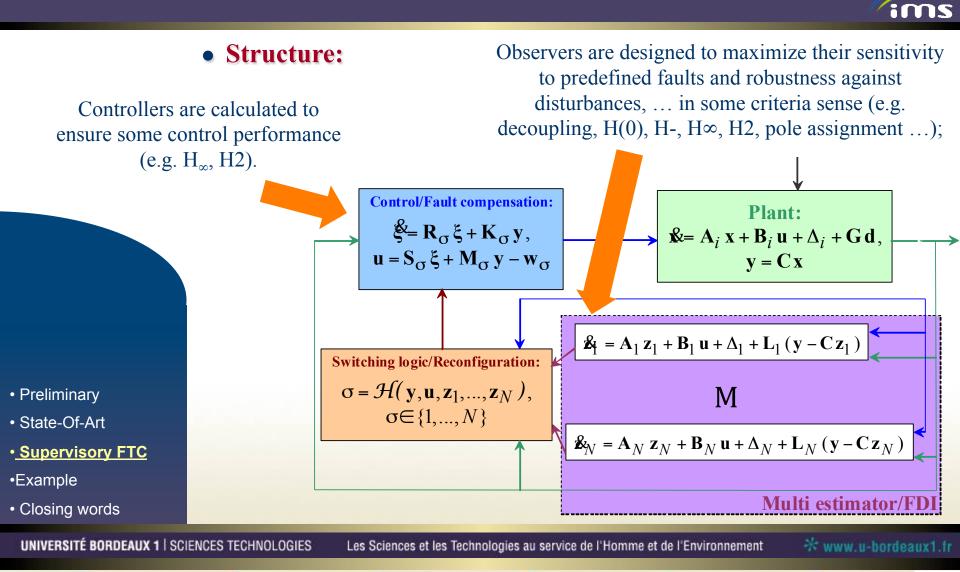






Supervisory FTC with mutual perf. 6/11





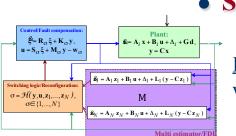
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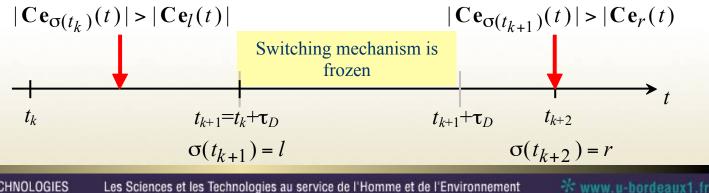
• Structure: Optimality of the subsystems does not imply the same property for the whole system.

Focus on dwell-time conditions: a mutual performance optimization can be done with minimization of the dwell-time value under FDI / control perf. constraints

Switching logic:

$$\begin{split} t_0 &= 0, t_{k+1} = \arg\inf_{t \ge t_k + \tau_D} \left\{ h \left| Ce_{\sigma(t_k)}(t) \right| > \left| Ce_j(t) \right|, j = 1, \dots, N, j \neq \sigma(t_k) \right\}, k \ge 0 \\ \sigma(t_k) &= \arg\min_{1 \le j \le N} \left| Ce_j(t_k) \right|, k \ge 0 \\ \sigma(t) &= \sigma(t_k) \quad \text{for all} \quad t_k \le t < t_{k+1}, k \ge 0 \end{split}$$

where $t_k, k \ge 0$ are instants of switches and $\tau_D > 0$ is the dwell-time constant.





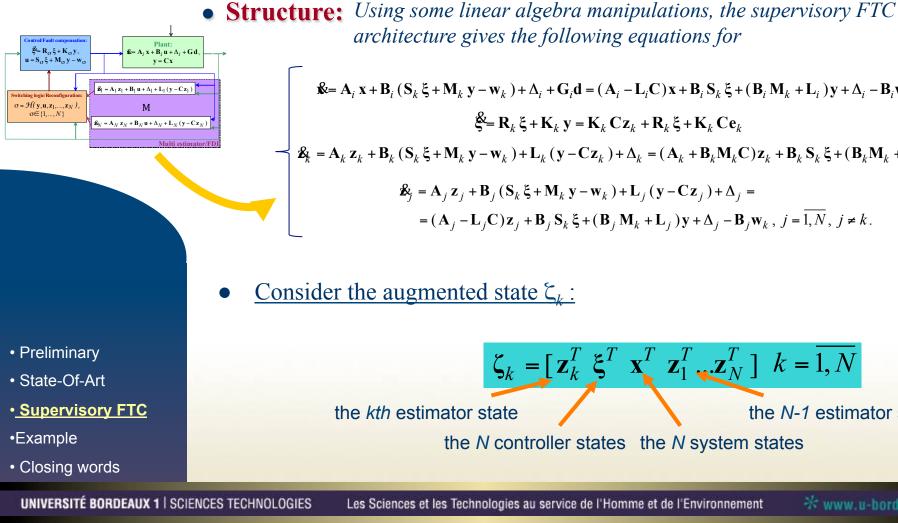




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$$\mathbf{\hat{k}} = \mathbf{A}_{i} \mathbf{x} + \mathbf{B}_{i} (\mathbf{S}_{k} \boldsymbol{\xi} + \mathbf{M}_{k} \mathbf{y} - \mathbf{w}_{k}) + \Delta_{i} + \mathbf{G}_{i} \mathbf{d} = (\mathbf{A}_{i} - \mathbf{L}_{i} \mathbf{C}) \mathbf{x} + \mathbf{B}_{i} \mathbf{S}_{k} \boldsymbol{\xi} + (\mathbf{B}_{i} \mathbf{M}_{k} + \mathbf{L}_{i}) \mathbf{y} + \Delta_{i} - \mathbf{B}_{i} \mathbf{w}_{k} + \mathbf{G}_{i} \mathbf{d}$$

$$\mathbf{\hat{\xi}} = \mathbf{R}_{k} \boldsymbol{\xi} + \mathbf{K}_{k} \mathbf{y} = \mathbf{K}_{k} \mathbf{C} \mathbf{z}_{k} + \mathbf{R}_{k} \boldsymbol{\xi} + \mathbf{K}_{k} \mathbf{C} \mathbf{e}_{k}$$

$$\mathbf{\hat{k}}_{k} = \mathbf{A}_{k} \mathbf{z}_{k} + \mathbf{B}_{k} (\mathbf{S}_{k} \boldsymbol{\xi} + \mathbf{M}_{k} \mathbf{y} - \mathbf{w}_{k}) + \mathbf{L}_{k} (\mathbf{y} - \mathbf{C} \mathbf{z}_{k}) + \Delta_{k} = (\mathbf{A}_{k} + \mathbf{B}_{k} \mathbf{M}_{k} \mathbf{C}) \mathbf{z}_{k} + \mathbf{B}_{k} \mathbf{S}_{k} \boldsymbol{\xi} + (\mathbf{B}_{k} \mathbf{M}_{k} + \mathbf{L}_{k}) \mathbf{C} \mathbf{e}_{k}$$

$$\mathbf{\hat{k}}_{j} = \mathbf{A}_{j} \mathbf{z}_{j} + \mathbf{B}_{j} (\mathbf{S}_{k} \boldsymbol{\xi} + \mathbf{M}_{k} \mathbf{y} - \mathbf{w}_{k}) + \mathbf{L}_{j} (\mathbf{y} - \mathbf{C} \mathbf{z}_{j}) + \Delta_{j} =$$

$$= (\mathbf{A}_{j} - \mathbf{L}_{j} \mathbf{C}) \mathbf{z}_{j} + \mathbf{B}_{j} \mathbf{S}_{k} \boldsymbol{\xi} + (\mathbf{B}_{j} \mathbf{M}_{k} + \mathbf{L}_{j}) \mathbf{y} + \Delta_{j} - \mathbf{B}_{j} \mathbf{w}_{k}, \ j = \overline{1, N}, \ j \neq k.$$

 $\boldsymbol{\zeta}_{k} = \begin{bmatrix} \mathbf{z}_{k}^{T} & \boldsymbol{\zeta}^{T} & \mathbf{x}_{1}^{T} & \mathbf{z}_{1}^{T} \dots \mathbf{z}_{N}^{T} \end{bmatrix} \quad k = 1, N$

Consider the augmented state ζ_k :

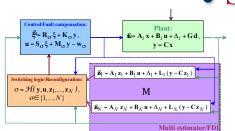
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the N-1 estimator states



Supervisory FTC with mutual perf. 9/11





• **Structure:** (Efimov, Cieslak, Henry, 2012)

A slow dynamic involves a long dwell-time

$$\boldsymbol{\xi}_{k}^{\boldsymbol{k}} = \mathbf{W}_{k,i} \,\boldsymbol{\zeta}_{k} + \mathbf{V}_{k,i} \,\mathbf{C} \,\mathbf{e}_{k} + \boldsymbol{\iota}_{k,i} + \boldsymbol{G}_{i}^{\boldsymbol{0}} \mathbf{d} \tag{1}$$

where the matrix $\mathbf{W}_{k,i}$ being left block triangular with the blocks on the main diagonal \mathbf{H}_k , $\mathbf{A}_i - \mathbf{L}_i \mathbf{C}$, $\mathbf{A}_1 - \mathbf{L}_1 \mathbf{C}$, $\mathbf{A}_N - \mathbf{L}_N \mathbf{C}$

Eq. (1) is stable.

- <u>Define the augmented state</u>: $\boldsymbol{\Psi} = [\boldsymbol{\xi}^T \ \mathbf{x}^T \ \mathbf{z}_1^T ... \mathbf{z}_N^T]$ Then, there exist permutation matrices \mathbf{T}_i so that $\boldsymbol{\Psi} = \mathbf{T}_i \boldsymbol{\zeta}_i$
- Owing the standard results on dwell-time switched systems stability (Liberzon, 2003; Morse, 1995; Xie et al., 2001; Efimov et al., 2008) the value of τ_D be taken to satisfy: $\tau_D = \max_{1 \le j \le N} \{-\alpha_{j,i}^{-1} \ln(\lambda \beta_{j,i}^{-1})\}$ minimal (in norm real part) of

$$\beta_{j,i} = \sup_{t \ge 0} |\exp(\mathbf{T}_j \mathbf{W}_{j,i} \mathbf{T}_j^{-1} t)|$$

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eigenvalues of the matrix \mathbf{W}_{ii}

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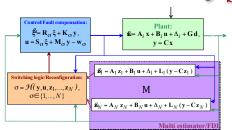
Closing words





Supervisory FTC with mutual perf. 10/11





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• Stability theorem: (Efimov, Cieslak, Henry, 2012)

Let i(t)=const for all $t \ge 0$. Then there exists τ_D such that for any $\psi(0)$ and finite ||d||

$$|\psi(t)| \le v_i e^{-\mu_i t/\tau_D} |\psi(0)| + v_i \{ ||\delta||_{[0,t)} + ||d||_{[0,t)} \} + \varpi_i \max_{1 \le i \le N} |\Delta_i|$$

for all t ≥0 and some parameters $v_i > 0$, $\mu_i > 0$, $\varpi_i > 0$ and $v_i > 0$ where

$$\boldsymbol{\delta}(t) = \begin{cases} \mathbf{C}[\mathbf{e}_i(t) - \mathbf{e}_{\sigma(t_k)}(t)] & \text{if } t \in [t_k, t_k + \tau_D) \land |\mathbf{C}\mathbf{e}_i(t)| < |\mathbf{C}\mathbf{e}_{\sigma(t_k)}(t)|; \\ 0 & \text{otherwise.} \end{cases}$$

minimum admissible time between two consecutive faults

• **Corollary:** (Efimov, Cieslak, Henry, 2012)

Let T_{r+1} - $T_r \ge T_D$ for all $r \ge 0$. Then there exist T_D and τ_D such that for any $\psi(0)$ and finite ||d||:

 $|\Psi(t)| \leq \Psi e^{-\beta \delta / T_D} |\Psi(0)| + \Psi ||\delta||_{[0,t)} + ||d||_{[0,t)} + W max_{1 \leq i \leq N} |\Delta_i|$

for all t ≥ 0 and some parameters $\widetilde{v}_i > 0$, $\widetilde{\mu}_i > 0$, $\widetilde{\varpi}_i > 0$ and $\widetilde{\upsilon}_i > 0$.

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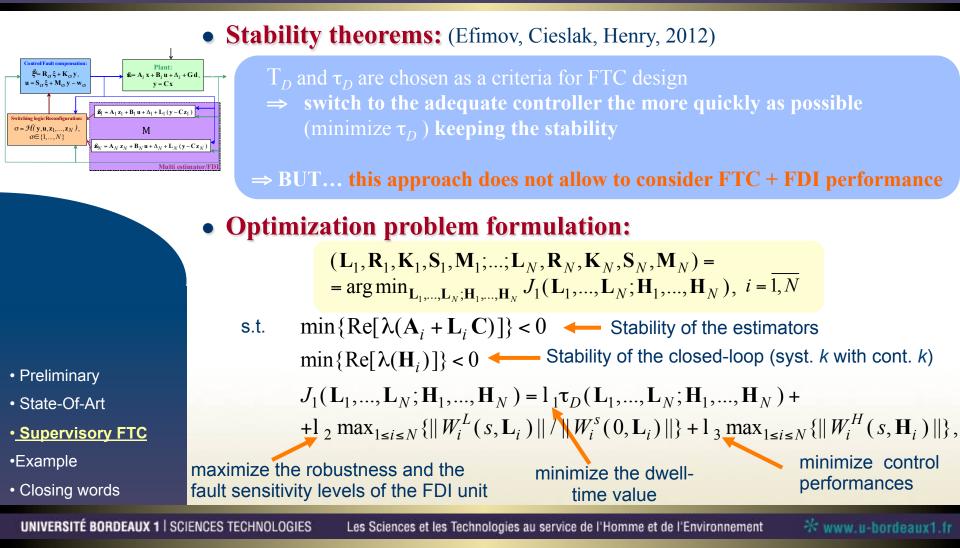




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Academic illustration 1/4



• Example: F-8 aircraft model (Zhang & Jiang, 2003)

$$\mathbf{A}_{1} = \begin{bmatrix} -3.598 & 0.1968 & -35.18 & 0 \\ -0.0377 & -0.3576 & 5.884 & 0 \\ 0.0688 & -0.9957 & -0.2163 & 0.0733 \\ 0.9947 & 0.1027 & 0 & 0 \end{bmatrix} \quad \mathbf{B}_{1} = \begin{bmatrix} 14.65 & 8.79 \\ 0.2179 & 0.1307 \\ -0.0054 & -0.0032 \\ 0 & 0 \end{bmatrix} \quad \mathbf{C} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 1 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad \mathbf{L}_{1} = \mathbf{C} \begin{bmatrix} 0$$

Considered faults: stuck actuators $A_2 = A_2$

$$\mathbf{A}_2 = \mathbf{A}_3 = \mathbf{A}_1$$

$$\mathbf{B}_{2} = \begin{bmatrix} 14.65 & 0 \\ 0.2179 & 0 \\ -0.0054 & 0 \\ 0 & 0 \end{bmatrix} \quad \Delta_{2} = \begin{bmatrix} 8.79 \\ 0.1307 \\ -0.0032 \\ 0 \end{bmatrix} \alpha_{2} \qquad \mathbf{B}_{3} = \begin{bmatrix} 0 & 8.79 \\ 0 & 0.1307 \\ 0 & -0.0032 \\ 0 & 0 \end{bmatrix} \quad \Delta_{2} = \begin{bmatrix} 8.79 \\ 0.1307 \\ -0.0032 \\ 0 \end{bmatrix} \alpha_{2}$$

Three distinguished operating modes *N***=**1,2,3,

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• **Example:** F-8 aircraft model (Zhang & Jiang, 2003) The nonlinear optimization problem is solved using a gridding approach 20 0.35 0 Solution (optimal) : 18 b 0.3 o 0.3 0.35 0.35 0.35 16 4 0.65 14 55 0 O. $\lambda(\mathbf{R}_i) \approx -4$. တ 12 Observers 10 0.4 0.4 8 Preliminary 6 0.4 0.4 State-Of-Art 0.4 n 4 4 5 Admissible H∞ 0 Supervisory FTC 0.45 performance zone 0.45 2 0.5 0.5 0.5 0.55 0.55 for control and FDI 12 16 2 4 6 10 14 18 20 8 Closing words Controls

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•Example

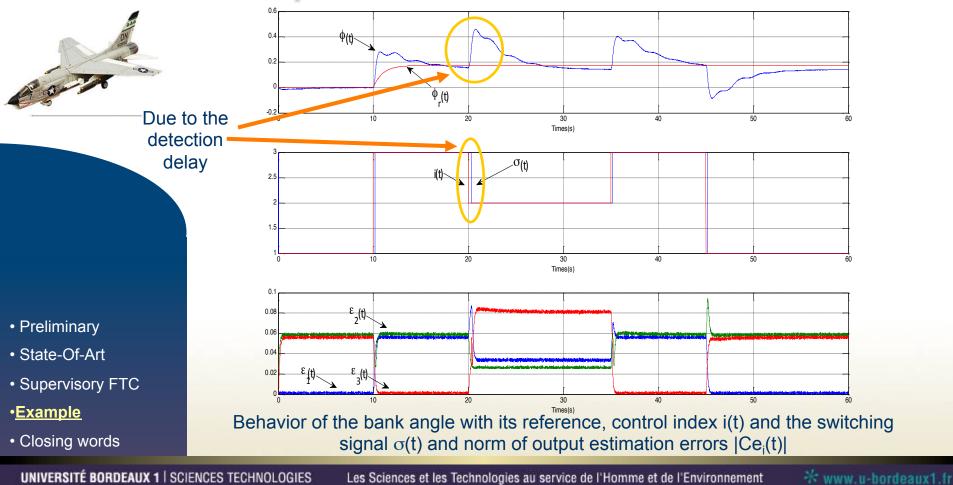




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• Example: F-8 aircraft model (Zhang & Jiang, 2003)



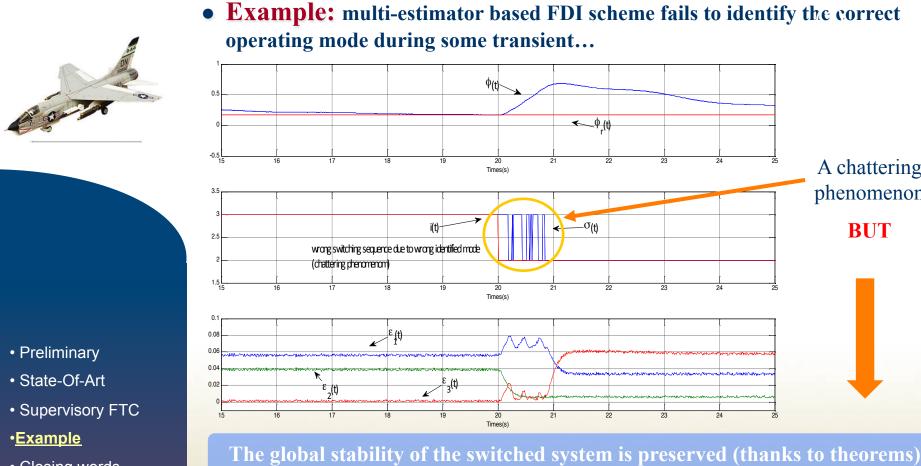
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A chattering

phenomenon

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Supervisory FTC

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- Advantages of this type of supervisory FTC approach
 - Formal stability proofs are established for the overall FTC scheme taking into account all units and bounded disturbances.
 - The method allows to design both the FDI and FTC units taking into account their coupling via an optimization problem.
 - It is proved that the global stability of the control law is preserved even if the FDI scheme fails to identify the correct fault.
- **Enhancement** (already developped in Cieslak, Efimov, Henry, Safeprocess'12)



- No management of undesirable transient behaviors due to switches Bumpless scheme is added -> Enhancement of dwell-time
- FT controllers have common state
- Overall stability is proved with no common state and/or different controller state dimension.

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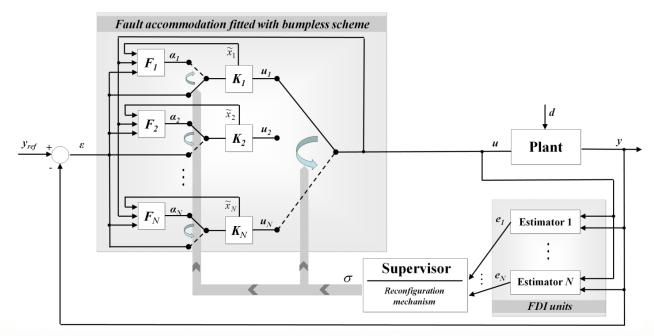


Closing words 2/4



• Enhancement: (Cieslak, Efimov, Henry, Safeprocess'12)

Meeting S3, Paris, CNAM - January, 18, 2013



- Preliminary
- State-Of-Art
- Supervisory FTC
- •Example

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<u>Closing words</u>

where F_i , i=1, ..., N are static gain to be designed such that the following quadratic criterion is minimized:

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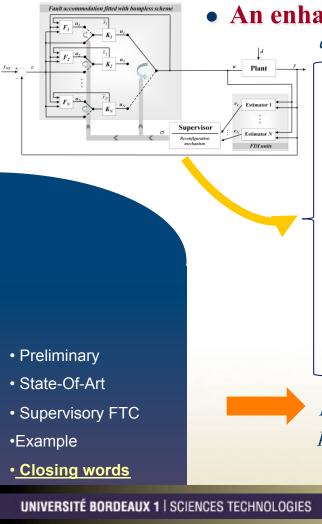
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Closing words 3/4





An enhancement: Using some linear algebra, the supervisory FTC architecture gives the following equations for j = 1, ..., N, $j \neq k$ $\vec{x} = A_i x + B_i (\widetilde{C}_k \widetilde{x}_k + \widetilde{D}_k y_{ref} - \widetilde{D}_k y) + \Delta_i + G_i d$ $= (A_i - L_iC)x + B_i\widetilde{C}_k\widetilde{x}_k + (L_i - B_i\widetilde{D}_k)Cz_k + (L_i - B_i\widetilde{D}_k)Ce_k + B_i\widetilde{D}_ky_{rot} + \Delta_i + G_id - B_i\widetilde{D}_kn$ $\dot{\widetilde{x}}_{k} = \widetilde{A}_{k}\widetilde{x}_{k} + \widetilde{B}_{k}y_{ref} - \widetilde{B}_{k}y = \widetilde{A}_{k}\widetilde{x}_{k} - \widetilde{B}_{k}Cz_{k} - \widetilde{B}_{k}Ce_{k} + \widetilde{B}_{k}y_{ref} - \widetilde{B}_{k}n$ $\dot{z}_{k} = A_{k} z_{k} + B_{k} (\widetilde{C}_{k} \widetilde{x}_{k} + \widetilde{D}_{k} y_{ref} - \widetilde{D}_{k} y) + \Delta_{k} + L_{k} (y - C z_{k})$ $= B_k \widetilde{C}_k \widetilde{x}_k + (A_k - B_k \widetilde{D}_k C) z_k + (L_k - B_k \widetilde{D}_k) C e_k + B_k \widetilde{D}_k y_{ref} + \Delta_k + (L_k - B_k \widetilde{D}_k) n$ $\dot{z}_{i} = A_{i} z_{i} + B_{i} (\widetilde{C}_{k} \widetilde{x}_{k} + \widetilde{D}_{k} y_{ref} - \widetilde{D}_{k} y) + \Delta_{i} + L_{i} (y - C z_{i})$ $=B_{i}\widetilde{C}_{k}\widetilde{x}_{k}+(L_{i}-B_{j}\widetilde{D}_{k})Cz_{k}+(A_{i}-L_{i}C)z_{i}+(L_{i}-B_{j}\widetilde{D}_{k})Ce_{k}+B_{j}\widetilde{D}_{k}y_{ref}+\Delta_{i}+(L_{i}-B_{j}\widetilde{D}_{k})n$ $\dot{\widetilde{x}}_{j} = \widetilde{A}_{j}\widetilde{x}_{j} + \widetilde{B}_{j}F_{j} \left(\begin{array}{c} \overline{x}_{j} \\ \widetilde{C}_{k}\widetilde{x}_{k} + \widetilde{D}_{k}y_{ref} - \widetilde{D}_{k}y \\ y_{ref} - y \end{array} \right)$ $=\widetilde{B}_{i}F_{2\,i}\widetilde{C}_{k}\widetilde{x}_{k}-\widetilde{B}_{i}\mathcal{N}Cz_{k}+(\widetilde{A}_{i}+\widetilde{B}_{i}F_{1\,i})\widetilde{x}_{i}-\widetilde{B}_{i}\mathcal{N}Ce_{k}+\widetilde{B}_{j}\mathcal{N}y_{ref}-\widetilde{B}_{j}\mathcal{N}n$ Introducing $\varsigma_k = [z_k^T \ \widetilde{x}_k^T \ x^T \ z_1^T \cdots z_N^T \ \widetilde{x}_1^T \cdots \widetilde{x}_N^T]^T$ such that $z_1^T \cdots z_N^T$ and parts $\widetilde{x}_k^T \circ n \widetilde{ot}_N^T \widetilde{c}$ ontained and respectively, $\widetilde{x}_k^T v \widetilde{c}$ have: $\dot{\varsigma}_{k} = W_{ki}\varsigma_{k} + V_{ki}Ce_{k} + \iota_{ki} + \widetilde{G}_{i}d + \overline{G}_{i}y_{ref} + \overline{G}_{i}n$

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Closing words 4/4



• And now... the perspectives:

- Identification and elaboration of solutions for tackling static and dynamic nonlinearities in supervisory control context by means of anti-windup solutions and/or Linear Parameter Varying (LPV) tools.
- The optimization problem formulation could be enhanced by formulating an optimization problem based on **Linear Matrix Inequalities** (LMI) framework. It offers an appealing context to manage the different specification trade-offs in terms of noise attenuation, sensitivity, tracking,.

- Preliminary
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- <u>Closing words</u>

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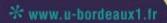


Thank you for your attention

Tolérance aux fautes par une approche superviseur: gestion du couplage FDI-FTC

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